

## SNA321H/358H/324H 1MHZ High Voltage Bipolar Opamp

#### Features

- Single-Supply Operation: +3V ~ +36V
- Dual-Supply Operation: ±1.5V ~ ±18V
- Gain-Bandwidth Product: 1MHz (Typ)
- Low Input Bias Current: 20nA (Typ)
- Low Offset Voltage: 5mV (Max)
- Quiescent Current: 250µA per Amplifier (Typ)
- Input Common Mode Voltage Range Includes
  Ground
- Large Outpu Voltage Swing: 0V to V<sub>DD</sub>-1.5V
- Operating Temperature: -25°C ~ +85°C
- Small Package:
  - SNA321H Available in SOT23-5 Package
  - SNA358H Available in SOP-8 and MSOP-8 Packages
  - SNA324H Available in SOP-14 Package

#### **General Description**

The SNA321H/358H/324H family have a high gain-bandwidth product of 1MHz, a slew rate of  $0.2V/\mu$ s, and a quiescent current of  $250\mu$ A at 5V. The SNA321H/358H/324H family is designed to provide optimal performance in low voltage and low noise systems.

#### Applications

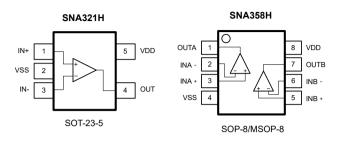
- Walkie-Talkie
- Battery Management Solution
- Transducer Amplifiers
- Summing Amplifiers
- Multivibrators
- Oscillators
- Switcching Telephone
- Portable Systems

#### **Ordering Information**

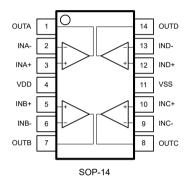
Model	Channel	Package	Ordering Number	Packing Option
SNA321H	Single	SOT-23-5	SNA321H00AB5	Tape and Reel,3000
SNA358H	SNA358H Dual	SOP-8	SNA358H00AA8	Tape and Reel,4000
SNASSON	Duai	MSOP-8	SNA358H00AM8	Tape and Reel,3000
SNA324H	Quad	SOP-14	SNA324H00AAF	Tape and Reel,2500



#### **Pin Assignment**



SNA324H



Pin	SOT-23-5	SOP-8/MSOP-8	SOP-14	Туре	Function
1	IN+	3/5	3/5/10/12	I	Noninverting input
2	VSS	4	11	—	Negative (lowest) power supply
3	IN-	2/6	2/6/9/13	I	Inverting input
4	OUT	1/7	1/7/8/14	0	Output
5	VDD	8	4	—	Positive (highest) power sup

#### **Absolute Maximum Ratings**

Parameter	Min	Мах	Unit
Power Supply Voltage (V <sub>DD</sub> to V <sub>SS</sub> )	-20	+20 or 40	V
Differential input voltage		40	V
Input Voltage	-0.3	40	V
Operating Temperature Range	-25	+85	°C
Storage Temperature Range	-65	150	°C

**Note:** Stresses greater than those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated under "Recommended Operating Conditions" is not implied. Exposure to "Absolute Maximum Ratings" for extended periods may affect device reliability.



## Contents

1	Electrical Characteristics	4
2	Typical Performance Characteristics	5
3	Application Note	7
4	Typical Application Circuits	9
5	Package Information	11
6	Revision History	15

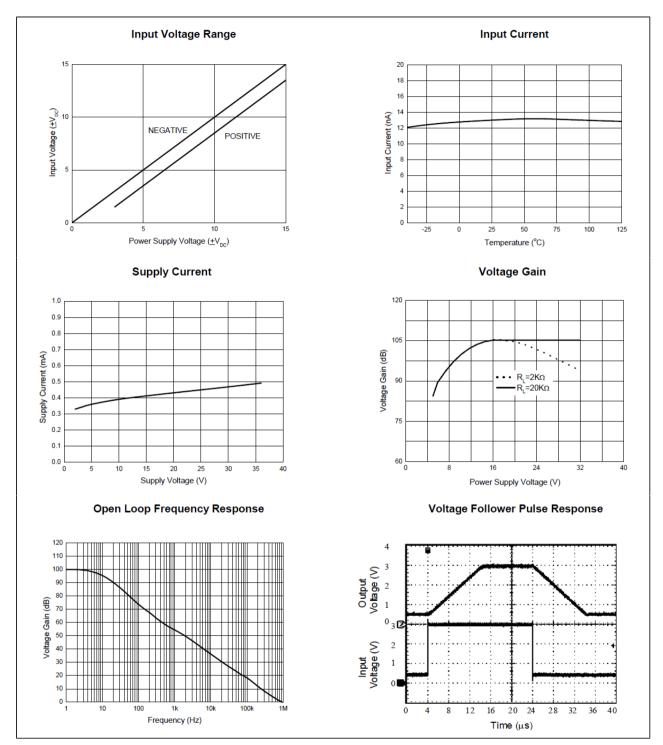


## **1** Electrical Characteristics

At  $V_{\text{DD}}$  = +5V,  $T_{\text{A}}$  = 25°C, unless otherwise noted.

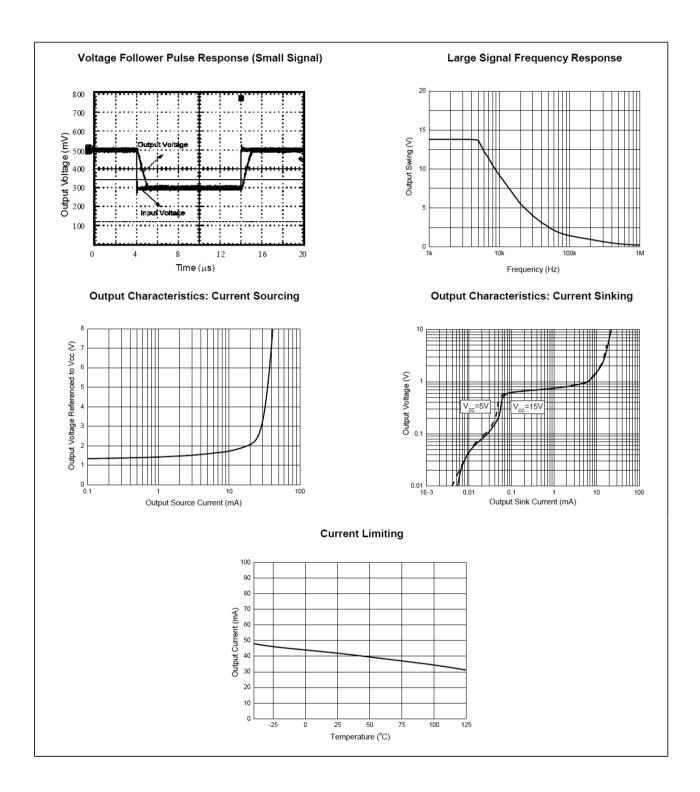
Parameter	Symbol	Conditions	Min	Тур	Мах	Unit
INPUT CHARACTERISTICS	!	*	·		•	<u>.</u>
Input Offset Voltage	V <sub>OS</sub>	$V_{CM} = V_{DD}/2$		0.4	5	mV
Input Bias Current	Ι <sub>Β</sub>			20		nA
Input Offset Current	I <sub>OS</sub>			5		nA
Common-Mode Voltage Range	V <sub>CM</sub>	V <sub>DD</sub> = 5.5V		-0.1~+4		V
Common-Mode Rejection Ratio	CMRR	$V_{CM} = 0V$ to $V_{DD}$ -1.5V	60	70		dB
Open-Loop Voltage Gain	A <sub>OL</sub>	$R_L = 5k\Omega$ , $V_O = 1V$ to 11V	85	100		dB
Input Offset Voltage Drift	ΔV <sub>OS</sub> /ΔT			7		µV/°C
OUTPUT CHARACTERISTICS				•		
	V <sub>OH</sub>	$R_L = 2k\Omega$		11		V
Output Voltage Swing from Boil	V <sub>OL</sub>	$R_L = 2k\Omega$		5	20	mV
Output Voltage Swing from Rail	V <sub>OH</sub>	R <sub>L</sub> = 10kΩ	1	12		V
	V <sub>OL</sub>	R <sub>L</sub> = 10kΩ		5	20	mV
0.4.4.0	I <sub>SOURCE</sub>	$R_L = 10\Omega$ to $V_{DD}/2$	1	40	60	mA
Output Current	I <sub>SINK</sub>			40	60	mA
POWER SUPPLY		•			•	
Operating Voltage Range			3		36	V
Power Supply Rejection Ratio	PSRR	V <sub>DD</sub> = +5V to +36V, V <sub>CM</sub> = +0.5V	70	100		dB
Quiescent Current / Amplifier	Ι <sub>Q</sub>	V <sub>DD</sub> = 36V, RL=∞		0.25	2.0	μA
DYNAMIC PERFORMANCE (C <sub>L</sub> = 1	00pF)			<u>.</u>	<u>.</u>	- <u>-</u>
Gain-Bandwidth Product	GBP			1		MHz
Slew Rate	SR	G = +1, 2V Output Step		0.2		V/µs





## 2 Typical Performance Characteristics







## 3 Application Note

### 3.1 Size

SNA321H/358H/324H family series op amps are unity-gain stable and suitable for a wide range of generalpurpose applications. The small footprints of the SNA321H/358H/324H family packages save space on printed circuit boards and enable the design of smaller electronic products.

### 3.2 Power Supply Bypassing and Board Layout

SNA321H/358H/324H family series operates from a single 3V to 36V supply or dual ±1.5V to ±18V supplies. For best performance, a  $0.1\mu$ F ceramic capacitor should be placed close to the  $V_{DD}$  pin in single supply operation. For dual supply operation, both  $V_{DD}$  and  $V_{SS}$  supplies should be bypassed to ground with separate  $0.1\mu$ F ceramic capacitors.

### 3.3 Low Supply Current

The low supply current (typical 250µA per channel) of SNA321H/358H/324H family will help to maximize battery life. They are ideal for battery powered systems.

### 3.4 Operating Voltage

SNA321H/358H/324H family operates under wide input supply voltage (3V to 36V). In addition, all temperature specifications apply from -25°C to +85°C. Most behavior remains unchanged throughout the full operating voltage range. These guarantees ensure operation throughout the single Li-Ion battery lifetime.

#### 3.5 Capacitive Load Tolerance

The SNA321H/358H/324H family is optimized for bandwidth and speed, not for driving capacitive loads. Output capacitance will create a pole in the amplifier's feedback path, leading to excessive peaking and potential oscillation. If dealing with load capacitance is a requirement of the application, the two strategies to consider are (1) using a small resistor in series with the amplifier's output and the load capacitance and (2) reducing the bandwidth of the amplifier's feedback loop by increasing the overall noise gain. Figure 3-1 shows a unity gain follower using the series resistor strategy. The resistor isolates the output from the capacitance and more importantly, creates a zero in the feedback path that compensates for the pole created by the output capacitance.

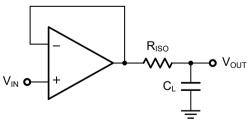


Figure 3-1 Indirectly Driving a Capacitive Load Using Isolation Resistor

The bigger the  $R_{ISO}$  resistor value, the more stable  $V_{OUT}$  will be. However, if there is a resistive load  $R_L$  in parallel with the capacitive load, a voltage divider (proportional to  $R_{ISO}/R_L$ ) is formed, this will result in a gain error.



The circuit in Figure 3-2 is an improvement to the one in Figure 3-1.  $R_F$  provides the DC accuracy by feedforward the V<sub>IN</sub> to R<sub>L</sub>.  $C_F$  and R<sub>ISO</sub> serve to counteract the loss of phase margin by feeding the high frequency component of the output signal back to the amplifier's inverting input, thereby preserving the phase margin in the overall feedback loop. Capacitive drive can be increased by increasing the value of C<sub>F</sub>. This in turn will slow down the pulse response.

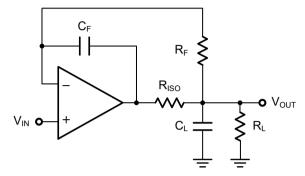


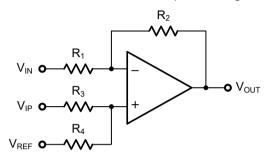
Figure 3-2 Indirectly Driving a Capacitive Load with DC Accuracy



## 4 **Typical Application Circuits**

#### 4.1 Differential amplifier

The differential amplifier allows the subtraction of two input voltages or cancellation of a signal common the two inputs. It is useful as a computational amplifier in making a differential to single-end conversion or in rejecting a common mode signal. Figure 4-1 shown the differential amplifier using SNA321H/358H/324H family.





$$V_{\text{OUT}} = \frac{(R_1 + R_2)}{(R_3 + R_4)} \frac{R_4}{R_1} V_{\text{IN}} - \frac{R_2}{R_1} V_{\text{IP}} + \frac{(R_1 + R_2)}{(R_3 + R_4)} \frac{R_3}{R_1} V_{\text{REF}}$$

If the resistor ratios are equal (i.e.  $R_1=R_3$  and  $R_2=R_4$ ), then

$$V_{\rm OUT} = \frac{R_2}{R_1} (V_{\rm IP} - V_{\rm IN}) + V_{\rm REF}$$

#### 4.2 Low Pass Active Filter

The low pass active filter is shown in Figure 4-2. The DC gain is defined by  $-R_2/R_1$ . The filter has a -20dB/ decade roll-off after its corner frequency  $f_c=1/(2\pi R_3 C_1)$ .

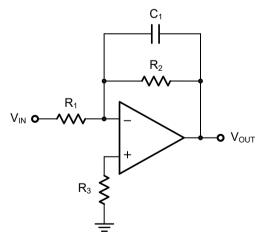


Figure 4-2 Low Pass Active Filter



## 4.3 Instrumentation Amplifier

The triple SNA321H/358H/324H family can be used to build a three-op-amp instrumentation amplifier as shown in Figure 4-3. The amplifier in Figure 4-3 is a high input impedance differential amplifier with gain of  $R_2/R_1$ . The two differential voltage followers assure the high input impedance of the amplifier.

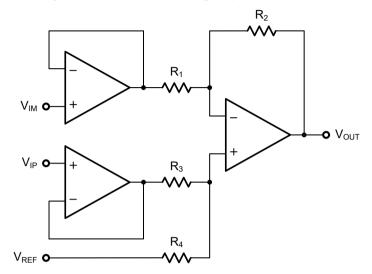
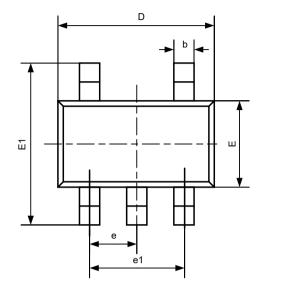


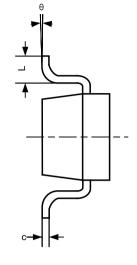
Figure 4-3 Instrument Amplifier

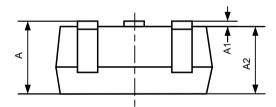


# 5 Package Information

### 5.1 SOT-23-5



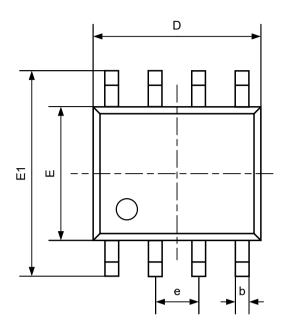


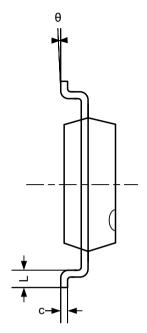


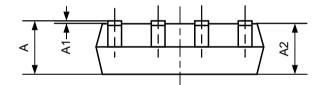
Symbol	Dimensions	in Millimeters	Dimension	s in Inches
Symbol	Min	Max	Min	Max
A	1.050	1.250	0.041	0.049
A1	0.000	0.100	0.000	0.004
A2	1.050	1.150	0.041	0.045
b	0.300	0.500	0.012	0.020
с	0.100	0.200	0.004	0.008
D	2.820	3.020	0.111	0.119
E	1.500	1.700	0.059	0.067
E1	2.650	2.950	0.104	0.116
е	0.950	) BSC	0.037 BSC	
e1	1.900	) BSC	0.075 BSC	
L	0.300	0.600	0.012	0.024
θ	0°	8°	0°	8°



## 5.2 SOP-8



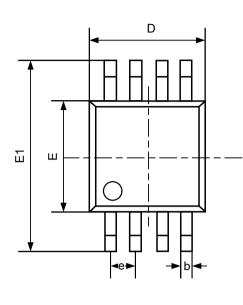


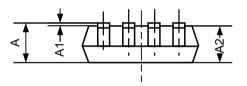


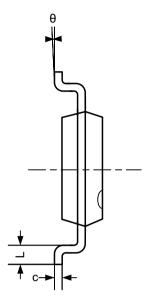
Symbol	Dimensions	n Millimeters	Dimensions in Inches	
Symbol	Min	Мах	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.350	1.550	0.053	0.061
b	0.330	0.510	0.013	0.020
C	0.170	0.250	0.006	0.010
D	4.700	5.100	0.185	0.200
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
е	1.270 BSC 0.050 B		BSC	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°



## 5.3 MSOP-8



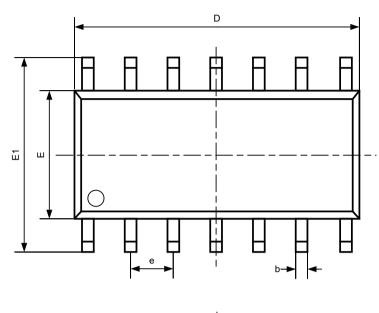


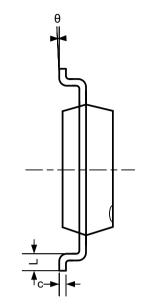


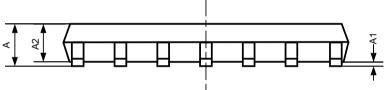
Symbol	Dimensions	Dimensions in Millimeters		s in Inches
Symbol	Min	Max	Min	Мах
Α	0.820	1.100	0.032	0.043
A1	0.020	0.150	0.001	0.006
A2	0.750	0.950	0.030	0.037
b	0.250	0.380	0.010	0.015
С	0.090	0.230	0.004	0.009
D	2.900	3.100	0.114	0.122
E	2.900	3.100	0.114	0.122
E1	4.750	5.050	0.187	0.199
е	0.650 BSC 0.026 BSC		BSC	
L	0.400	0.800	0.016	0.031
θ	0°	6°	0°	6°



## 5.4 SOP-14







Symbol	Dimensions i	n Millimeters	<b>Dimensions in Inches</b>		
Symbol	Min	Мах	Min	Max	
A	1.350	1.750	0.053	0.069	
A1	0.100	0.250	0.004	0.010	
A2	1.250	1.650	0.049	0.065	
b	0.360	0.490	0.014	0.019	
C	0.130	0.250	0.005	0.010	
D	8.530	8.730	0.336	0.344	
E	3.800	4.000	0.150	0.157	
E1	5.800	6.200	0.228	0.244	
е	1.270 BSC 0.050 BSC		BSC		
L	0.450	0.800	0.018	0.032	
θ	0°	8°	0°	8°	



#### 6 Revision History

Version	Date	Description
0.1	2022/07/12	Initial release

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